Multiuser Detection with Particle Filtering *

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ABSTRACT

In this paper, we apply particle filtering to multiuser detection (MUD) in synchronous CDMA systems with perfect channel state information (CSI). To apply particle filtering to MUD, first we find a factored representation of the posterior distribution. We show that the whitened matched filter (WMF) output allows such representation. No system approximations are made and the method is "soft" in nature. As a result, a near optimum performance is observed in simulations both in the equal power case and the near-far case. Since the computational complexity of the method is not exponential with the number of users, the method may have potential for industrial application.

1 Introduction

When MUD was introduced in the eighties, it quickly received a great deal of attention due to its potential for reducing the effects of multiple access interference (MAI) and thereby for increasing the capacity of CDMA systems. Since optimum maximum likelihood MUD is exponential in complexity, numerous approximate detectors were developed for MUD. In general, successive/iterative detectors [3], such as the decorrelating decision-feedback detector, perform better than their corresponding "one-pass" detectors like, for example, the decorrelating detector. These successive/iterative detectors usually form interim hard decisions for later stages and are, thus, prone to error propagation. Consequently, the performance of these conventional detectors is not near optimum.

In [10], a Markov chain Monte Carlo method, the Gibbs sampler, was used for CDMA MUD. It was implemented in a Bayesian framework, and it was demonstrated that it could provide near optimum performance. It is "soft" in nature, that is, the method allows for exchange of "extrinsic" information in an iterative (turbo) joint MUD and channel decoding. However, the Gibbs sampler has inherent drawbacks. It is hard to determine when the underlying Markov chain converges, and sometimes, the Gibbs sampler gets stuck at a local optimum. As a result, when we experimented with it on MUD, there was an error floor in its performance.

In this paper, we aim to develop a Bayesian-based solution that can outperform conventional detectors and overcome the shortcomings of the Gibbs sampler. To that end we would like to keep the capability of providing "extrinsic" information while performing MUD. We propose a particle filtering approach, a methodology that has reemerged recently in the fields of engineering [4], econometrics [8], and statistics [5]. However, to apply particle filtering, one needs to find a dynamic state space model representation of the system which in our problem is not obvious. We show that the whitened matched filter (WMF) output, which is used in decorrelating decision feedback detectors, allows such a representation. Note that our work is different from that in [6], where the binary data of all users in a symbol interval are considered as a super symbol. As a result, the sample space grows exponentially with the number of users. By contrast, our algorithm samples one user at a time in the binary space and can handle large number of users.

2 System description

Consider a synchronous CDMA system with a chip rate (processing gain) C and K users. Let T denote the symbol duration and $s_k(t)$ the normalized deterministic signature waveform assigned to the kth user. Here $t \in$ [0,T] and $k \in \{1,\ldots,K\}$. Let $b_k \in \{-1,+1\}$ be the bit transmitted by the kth user, a_k the channel state information of the kth user, and $\sigma n(t)$ the received zero mean complex white Gaussian noise with variance σ^2 . We can express the received signal y(t) as

$$y(t) = \sum_{k=1}^{K} a_k b_k s_k(t) + \sigma n(t) \qquad t \in [0, T].$$
(1)

The received signal is a superposition of K antipodally modulated synchronous signature waveforms plus noise. The cross-correlation between the signature waveforms

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of the ith and the jth users is defined according to

$$\mathbf{R}_{ij} = \langle s_i, s_j \rangle = \int_0^T s_i(t) s_j(t) dt \tag{2}$$

where \mathbf{R}_{ij} is the *ij*th element of the cross-correlation matrix \mathbf{R} .

In CDMA systems, we often work with the matched filter output,

$$y_k = \langle y(t), s_k(t) \rangle = \int_0^T y(t) s_k(t) dt.$$
 (3)

The set of matched filter outputs $\mathbf{y} = [y_1 \ y_2 \ \cdots \ y_K]^{\mathsf{T}}$ (^T stands for vector or matrix transposition) can be represented in a vector-matrix form according to

$$\mathbf{y} = \mathbf{R}\mathbf{A}\mathbf{b} + \mathbf{n} \tag{4}$$

where $\mathbf{A} = \text{diag}\{a_1, \dots, a_K\}$ is the diagonal matrix of the channel state information, $\mathbf{b} = [b_1 \ b_2 \ \dots \ b_K]^{\mathsf{T}}$ is the user symbol vector, and \mathbf{n} is a complex-valued Gaussian vector with independent real and imaginary components and covariance matrix equal to $\sigma^2 \mathbf{R}$.

The cross-correlation matrix is positive definite, and we can factor it by Cholesky decomposition. There exists a unique lower triangular matrix \mathbf{F} such that $\mathbf{R} = \mathbf{F}^{\mathsf{T}}\mathbf{F}$. If we apply $\mathbf{F}^{-\mathsf{T}} = (\mathbf{F}^{\mathsf{T}})^{-1}$ to the matched filter output, we obtain

$$\bar{\mathbf{y}} = (\mathbf{F}^{\mathsf{T}})^{-1}\mathbf{y} = \mathbf{F}^{-\mathsf{T}}\mathbf{F}^{\mathsf{T}}\mathbf{F}\mathbf{A}\mathbf{b} + \mathbf{F}^{-\mathsf{T}}\mathbf{n} = \mathbf{F}\mathbf{A}\mathbf{b} + \bar{\mathbf{n}},$$
(5)

or equivalently

$$\bar{\mathbf{y}} = \mathbf{FBa} + \bar{\mathbf{n}} \tag{6}$$

where $\mathbf{B} = \text{diag}\{b_1, b_2, \cdots, b_K\}$ is the user symbol matrix, and **a** is the $K \times 1$ vector of channel state information. The covariance matrix of $\bar{\mathbf{n}}$ becomes

$$E\left(\bar{\mathbf{n}}\bar{\mathbf{n}}^{\mathsf{T}}\right) = \sigma^{2}\mathbf{F}^{\mathsf{T}}\mathbf{R}\mathbf{F}^{-1} = \sigma^{2}\mathbf{I}$$
(7)

where I is the identity matrix. Since the noise becomes i.i.d. white Gaussian, $\bar{\mathbf{y}}$ is called the whitened matched filter(WMF) output. The expression in (6) can be written in component-wise form as

$$\bar{y}_k = \sum_{l=1}^{K} F_{k,l} a_l b_l + \bar{n}_k, \quad k = 1, 2, \cdots, K.$$
 (8)

In the sequel, let $\bar{y}_{1:k} = {\bar{y}_1, \dots, \bar{y}_k}$, and define $b_{1:k}$ and $a_{1:k}$ similarly. The objective is to detect $b_{1:k}$ given $\bar{y}_{1:k}$.

3 Particle filtering MUD with perfect CSI

All the prior information and information from the data is combined in the posterior distribution $p(b_{1:K}|\bar{y}_{1:K})$. Direct evaluation of the posterior is impossible for large K due to its high dimensionality. An alternative is to obtain samples (*trajectories*) $\{b_{1:K}^{(j)}\}_{j=1}^{J}$ and weights $\{w_{1:K}^{(j)}\}_{j=1}^{J}$ associated with the trajectories, where j is the trajectory index, and using them, to compute desired estimates. For example, we can approximate the expected value of any function of $b_{1:K}$, $f(b_{1:K})$, by

$$E(f(b_{1:K})) \simeq \frac{1}{\sum_{j=1}^{J} w^{(j)}} \sum_{j=1}^{J} w^{(j)} f(b_{1:K}^{(j)}).$$
(9)

If the trajectories are drawn from the posterior distribution itself, all the trajectories have equal weights. Otherwise, if they are taken from a proposal importance function $\pi(b_{1:K}|\bar{y}_{1:K})$, the weights are evaluated according to

$$w^{(j)} = \frac{p(b_{1:K}^{(j)} | \bar{y}_{1:K})}{\pi(b_{1:K}^{(j)} | \bar{y}_{1:K})}, \quad \forall j.$$
(10)

By the law of large numbers, under certain conditions, the approximation in (9) reaches the true value when the number of samples J approaches infinity.

The posterior distribution can be factored according to

$$p(b_{1:k}|\bar{y}_{1:k}) \propto p(\bar{y}_k|b_{1:k})p(b_k)p(b_{1:k-1}|\bar{y}_{1:k-1})$$
(11)

and if we choose the proposal distribution in the form of $\pi(b_{1:k}|\bar{y}_{1:k}) = \pi(b_{1:k-1}|\bar{y}_{1:k-1})\pi(b_k|b_{1:k-1},\bar{y}_{1:k})$, we can form the trajectories recursively. Suppose that from previous iterations we have generated the trajectories $\{b_{1:k-1}^{(j)}\}_{j=1}^J$ from $\pi(b_{1:k-1}|\bar{y}_{1:k-1})$ with weights $\{w_{k-1}^{(j)}\}_{j=1}^J$. If we draw particles $b_k^{(j)}$ from the importance proposal distribution $\pi(b_k|b_{1:k-1}^{(j)},\bar{y}_{1:k}) = p(b_k|b_{1:k-1}^{(j)},\bar{y}_{1:k})$ and append them to $b_{1:k-1}^{(j)}$, the expended trajectory $b_{1:k}^{(j)}$ can be weighted with respect to $p(b_{1:k}|\bar{y}_{1:k})$ according to

$$\begin{split} w_{k}^{(j)} &= \frac{p(b_{1:k}^{(j)} | \bar{y}_{1:k})}{p(b_{k}^{(j)} | b_{1:k-1}^{(j)}, \bar{y}_{1:k}) \pi(b_{1:k-1}^{(j)} | \bar{y}_{1:k-1})} \\ &= \frac{p(\bar{y}_{k} | b_{1:k}^{(j)}) p(b_{k}^{(j)}) p(b_{1:k-1}^{(j)} | \bar{y}_{1:k-1})}{p(\bar{y}_{k} | \bar{y}_{1:k-1}) \frac{p(\bar{y}_{k} | b_{1:k-1}^{(j)} \bar{y}_{1:k-1})}{p(\bar{y}_{k} | b_{1:k-1}^{(j)} \bar{y}_{1:k-1})} \pi(b_{1:k-1}^{(j)} | \bar{y}_{1:k-1})} \\ &\propto w_{k-1}^{(j)} p(\bar{y}_{k} | b_{1:k-1}^{(j)}, \bar{y}_{1:k-1}) \\ &\propto w_{k-1}^{(j)} \sum_{b_{k}} p(\bar{y}_{k} | b_{k}, b_{1:k-1}^{(j)}) p(b_{k}) \end{split}$$
(12)

where $w_{k-1}^{(j)} = \frac{p(b_{1:k-1}^{(j)}|\bar{y}_{1:k-1})}{\pi(b_{1:k-1}^{(j)}|\bar{y}_{1:k-1})}$ is obtained from the previous iteration. In deriving the above equation, we utilized the fact that $p(b_k^{(j)}|b_{1:k-1}^{(j)}, \bar{y}_{1:k-1}) = p(b_k^{(j)})$, i.e., b_k is independent of other users and previous observations. We have also ignored the term $p(\bar{y}_k|\bar{y}_{1:k-1})$ because it is the same for all trajectories. The importance proposal distribution used here is referred to as optimal because it takes into account all the previous particles and all available observations, and as a result produces weights

with minimal variance conditional on $b_{1:k-1}^{(j)}$ and $\bar{y}_{1:k}$ [2]. the proposal distribution can be evaluated according to

$$\pi(b_{k}|b_{1:k-1}^{(j)}, \bar{y}_{1:k}) = p(b_{k}|b_{1:k-1}^{(j)}, \bar{y}_{1:k})$$

$$\propto p(\bar{y}_{k}|b_{k}, b_{1:k-1}^{(j)}, \bar{y}_{1:k-1})p(b_{k}|b_{1:k-1}, \bar{y}_{1:k-1})$$

$$= p(\bar{y}_{k}|b_{k}, b_{1:k-1}^{(j)})p(b_{k}).$$
(13)

Note that the weight is proportional to the sum of the proposal densities in (13). In our case, since $b_k \in \{+1, -1\}$, there are only two proposal densities to evaluate and the weight can be easily obtained.

In summary, the algorithm proceeds as follows:

For the iteration of the kth user and the jth trajectory,

- 1. Draw a particle $b_k^{(j)}$ from the proposal distribution (13).
- 2. Append $b_k^{(j)}$ to $b_{1:k-1}^{(j)}$ and obtain the new trajectory $b_{1:k}^{(j)}$.
- 3. Evaluate the weight of the jth trajectory using (12).

When the algorithm is completed with the last user K, we have trajectories and weights $\{(b_{1:K}^{(j)}, w_K^{(j)})\}_{j=1}^J$ that can approximate $p(b_{1:K}^{(j)} | \bar{y}_{1:K})$, the desired posterior distribution. This process of recursively obtaining the particles $b_k^{(j)}$ is called particle filtering.

From the generated particles and their weights, we can obtain various types of estimates. For example, the marginalized posterior distribution can be approximated by

$$p(b_k | \bar{y}_{1:K}) \simeq \frac{1}{\sum_{j=1}^J w_K^{(j)}} \sum_{j=1}^J w_K^{(j)} \delta(b_k - b_k^{(j)}) \qquad (14)$$

where $\delta(\cdot)$ is the Dirac delta function. If the adopted estimate of b_k is the one that maximizes the marginalized posterior distribution, we have

$$\hat{b}_{k} = \arg \max_{b_{k}} p(b_{k} | \bar{y}_{1:K})$$

$$\simeq \arg \max_{b_{k}} \frac{1}{\sum_{j=1}^{J} w_{K}^{(j)}} \sum_{j=1}^{J} w_{K}^{(j)} \delta(b_{k} - b_{k}^{(j)}). (15)$$

If $\mathbf{b}_k = [b_k^{(1)}, \cdots, b_k^{(J)}]^\mathsf{T}$ and $\mathbf{w}_K = [w_K^{(1)}, \cdots, w_K^{(J)}]^\mathsf{T}$, (15) can be simplified as

$$\hat{b} = \operatorname{sign}(\mathbf{b}_k^T \mathbf{w}_K).$$
 (16)

Note that "extrinsic" information can also be derived from (14).

An important issue of the particle filtering process is the need for resampling. Namely, after several steps, some weights of the samples become trivial and stop contributing to the overall estimates. In the literature of particle filtering, resampling is used so that samples with negligible weights are replaced by those from high distribution areas of the desired posterior distribution. There are many strategies for resampling, and we use the residual resampling procedure as described in [1].

The complexity of the algorithm is O(KJ), i.e., proportional to the product of the number of particles and number of users. If the number of particles is fixed, then the complexity is only linear with respect to the number of users.

4 Simulations

We simulated a synchronous CDMA system with K = 15 users with equal powers and a chip rate of C = 30. The spreading codes were generated randomly and the same spreading code was used in all experiments. Residual resampling was performed after every 5 users. The results of the performance comparison with other popular CDMA multiuser detectors are presented in Figure 1. The performance curves in the figure were obtained by averaging the Bit-Error Rates (BERs) of all 15 users.

The detectors used in the comparison include the three-stage successive cancellation detector with decorrelating first stage (3-stage) [9], the detector based on Gibbs sampling [10], and the decorrelating decision feedback detector (DDF) [3]. For the Gibbs sampler, we have experimented with two scenarios with different burn-in periods (the periods until convergence). In the first case, 100 samples were generated of which the first 50 samples were discarded (Gibbs-50). In the second case, 150 samples were drawn, and the first 100 samples were discarded (Gibbs-100). As a reference, we used the breadth-first tree-search algorithm [7], which is optimal, to provide a lower bound for the detectors.

From the results, we see that the particle filtering provides near-optimum performance. We used two particle filtering detectors, one with 50 particles for each user (PF-50) and another with 100 particles (PF-100). It appears that the performance gain by increasing the number of particles from 50 to 100 is only marginal.

The figure also shows that at high SNRs, the Gibbs sampler gets trapped at some local optimum and that it takes long time for the algorithm to converge to the global optimum. Consequently, with a limited burn-in period, the Gibbs sampler exhibits an error floor. In comparison, the particle filtering based detector does not have this problem.

We also investigated the performance of these detectors in a near-far scenario. In our experiment, the targeted user (the first user), had an SNR of 9 dB, and the signal strength of the remaining 14 users was identical. In comparison with the power of the targeted user, the power of the remaining users, E_b/E_1 , varied from -10dB to 10 dB. In Figure 2, we plotted the BER of the first user as a function of E_b/E_1 . It is clear that particle filtering almost always outperforms the 3-stage detector and although it performs worse than the Gibbs sampler with weaker interferers, it is more consistent than



Figure 1: Performance comparison of various detectors for C=30, K=15, and equal power with perfect CSI.



Figure 2: BER performance comparison of various detectors in a near-far scenario for C=30, K=15, and perfect CSI. The SNR of the first user was 9 dB.

the Gibbs sampler and it is near optimum in the range from -4 dB to 10dB. Note that during the simulation, the users are arranged in descending order according to their power. The ordering has some impact on the performance of particle filtering. The suboptimal performance from -10 dB to -6 dB may be caused by the ad-hoc resampling procedure or the limited number of particles. The performance improvement of the particle filtering based detector will be the objective of further research.

5 Conclusions

In this paper, we used the WMF output to derive a factored representation of the posterior distribution function for the MUD problem, and the representation allowed for application of particle filtering. The proposed method provides consistently better performance than several existing detectors which have been tested. The complexity of the method is linear in the number of users and therefore, it has great potential in practice. The method, however, needs further study which may lead to additional improvements of its performance. For example, a thorough examination of the detector's performance as a function of the number of particles and number of users is necessary, and other strategies for resampling should be explored.

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