

DERIVATIVE COMPRESSIVE SAMPLING WITH APPLICATION TO PHASE UNWRAPPING

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ABSTRACT

Reconstruction of multidimensional signals from the samples of their partial derivatives is known to be an important problem in imaging sciences, with its fields of application including optics, interferometry, computer vision, and remote sensing, just to name a few. Due to the nature of the derivative operator, the above reconstruction problem is generally regarded as ill-posed, which suggests the necessity of using some a priori constraints to render its solution unique and stable. The ill-posed nature of the problem, however, becomes much more conspicuous when the set of data derivatives occurs to be incomplete. In this case, a plausible solution to the problem seems to be provided by the theory of *compressive sampling*, which looks for solutions that fit the measurements on one hand, and have the sparsest possible representation in a predefined basis, on the other hand. One of the most important questions to be addressed in such a case would be of how much incomplete the data is allowed to be for the reconstruction to remain useful. With this question in mind, the present note proposes a way to augment the standard constraints of compressive sampling by additional constraints related to some natural properties of the partial derivatives. It is shown that the resulting scheme of *derivative compressive sampling* (DCS) is capable of reliably recovering the signals of interest from much fewer data samples as compared to the standard CS. As an example application, the problem of phase unwrapping is discussed.

1. INTRODUCTION

Numerous applications are known in which one is provided with the measurements of the gradient of a multidimensional signal, rather than of the signal itself. Central to such applications, therefore, appears the problem of reconstruction of signals from their partial derivatives subject to some a priori constraints (which could be either probabilistic or deterministic in nature) [1]. One of such applications, which has been chosen to exemplify the major contribution of this note, is the problem of *phase unwrapping*. Note that solving this problem is known to be a standard procedure in, e.g., optical and synthetic aperture radar (SAR) interferometry [2], stereo vision [1], blind deconvolution [3, 4], etc.

In order to specify the problem of phase unwrapping, let $F(x, y)$ be an arbitrary continuously differentiable function defined over a closed subset of the real plane \mathbb{R}^2 . If F happens to be the phase of a complex-valued function, it can only be measured in its *wrapped* form, i.e. *modulo* 2π . Formally, the process of phase wrapping can be represented by its associated operator $\mathbf{W} : \mathbb{R}^2 \rightarrow (-\pi, \pi]$. In this notation, the wrapped *principal* phase R is given as $R = \mathbf{W}[F]$. Specifi-

cally, the operator \mathbf{W} adds to F a piecewise-constant function $K : \mathbb{R}^2 \rightarrow \{2\pi k\}_{k \in \mathbb{Z}}$ resulting in $R = \mathbf{W}[F] = F + K$ that obeys [5]:

$$-\pi < \mathbf{W}[F(x, y)] \leq \pi, \quad \forall (x, y) \in \mathbb{R}^2. \quad (1)$$

In complex notation, the gradients of F and R can be defined as

$$\begin{aligned} \nabla F &= \frac{\partial F}{\partial x_1} \mathbf{i} + \frac{\partial F}{\partial x_2} \mathbf{j} \\ \nabla K &= \frac{\partial K}{\partial x_1} \mathbf{i} + \frac{\partial K}{\partial x_2} \mathbf{j}, \end{aligned} \quad (2)$$

where \mathbf{i} and \mathbf{j} denote the unit vectors associated with the x - and y -axis, respectively. Consequently, computing the gradient of the wrapped phase R using equations (2) yields

$$\nabla R = \nabla \mathbf{W}[F] = \nabla F + \nabla K. \quad (3)$$

Finally, applying the wrapping operator \mathbf{W} one more time to both sides of (3) results in

$$\mathbf{W}[\nabla \mathbf{W}[F]] = \mathbf{W}[\nabla R] = \nabla F + \nabla K + K'. \quad (4)$$

Due the property of operator \mathbf{W} to produce the values in interval $[-\pi, \pi]$, the term $K + K'$ vanishes as long as [2]

$$-\pi < \nabla F \leq \pi. \quad (5)$$

Therefore, as long as the condition (5) above holds, the gradient of the original phase F can be unambiguously recovered from the gradient of the corresponding principal phase R according to

$$\nabla F = \mathbf{W}[\nabla R]. \quad (6)$$

From the above considerations it follows that, if the condition (5) was known to hold then, given a measured R , an estimate \hat{F} of the original phase F could be obtained as a solution to the following optimization problem

$$\hat{F} = \arg \min_F \int \int \|\nabla F - \mathbf{W}[\nabla R]\|^2 dx dy, \quad (7)$$

which amounts to solving a Poisson equation subject to appropriate boundary conditions. Unfortunately, situations are rare in which the condition (5) can be a priori guaranteed. In this case, the estimate of ∇F as $\mathbf{W}[\nabla R]$ is contaminated by, so called, residuals, which cause the solution of (7) to be of little practical value.

In order to overcome the limitations of phase unwrapping inflicted by using the gradients estimated according to (6),

a magnitude of different approaches has been hitherto proposed [2]. In the current note, we introduce a different solution to the problem which is based on the concepts of the theory of compressive sampling [5, 6, 7, 8, 9, 10]. In particular, let $\Gamma \subset \mathbb{R}^2$ be a finite discrete subset over which the values of F need to be recovered. Let further Γ_0 denote a subset of those points in Γ at which the condition (5) is known to hold, and hence at which the gradient ∇F estimated according to (6) can be assumed to be errorless. (Note that the subset Γ_0 can be identified based on analysis of the field of residuals as detailed in [2]). Subsequently, we first recover the values of ∇F over the whole Γ from its *incomplete* measurements over Γ_0 , followed by estimating the original phase F using (6). Moreover, in addition to the standard constraints of compressive sampling, we propose to use the constraints stemming from the nature of the gradient as a potential field, viz.

$$\frac{\partial F(x,y)}{\partial x \partial y} = \frac{\partial F(x,y)}{\partial y \partial x}. \quad (8)$$

We will refer to the problem of reconstruction of F from $\{\nabla F(x,y)\}_{(x,y) \in \Gamma_0}$ as the problem of *derivative compressive sampling* (DCS), and show that using (8) allows considerably reducing the cardinality of Γ_0 , while preserving a predefined error rate.

The remainder of this paper is organized as follows. Section 2 provides an overview of compressive sampling, and shows how this theory can be used for solution of the problem of phase unwrapping. In Section 3, some necessary technical details are specified. The performance of the method is analyzed in Section 4, while Section 5 finalizes the paper with a discussion and an outline of our future research directions.

2. DERIVATIVE COMPRESSIVE SAMPLING

The theory of compressive sampling addresses the problem of *perfect* reconstruction of signals of interest from their *subcritically sampled* measurements [5, 6, 7, 8, 9, 10, 11]. In the case when incomplete measurements of the derivatives of the signals are available, the resulting reconstruction problem is referred to as derivative compressive sampling.

2.1 Basics of Compressive Sampling

The idea of compressive sampling was first formulated by D. Donoho [6] in the form of the, so called, generalized uncertainty principle. In this initial setup, a bandlimited signal $f(t) \in \mathbb{L}_2(\mathbb{R})$ is used for transmission over a channel, in which it ‘‘loses’’ its values on a subset T . Formally, one can define

$$r(t) = (\mathbf{I} - \mathbf{P}_T) f(t) + n(t), \quad (9)$$

where \mathbf{I} denotes the identity operator, $n(t)$ is observation noise, and \mathbf{P}_T denotes the spatial limiting operator of the form

$$\mathbf{P}_T f(t) = \begin{cases} f(t), & t \in T \\ 0, & \text{otherwise} \end{cases} \quad (10)$$

The second operator used in [6] is a band-limiting operator defined as given by

$$\mathbf{P}_\Omega f(t) \equiv \int_{\Omega} \hat{f}(\omega) e^{2\pi i \omega t} d\omega, \quad (11)$$

where $\hat{f}(\omega)$ denotes the Fourier transform of $f(t)$.

The main goal of compressive sampling is to reconstruct the transmitted signal f from the noisy received signal r . The possibility of such a recovery is assured by Theorems 2 & 4 in [6] asserting that if $|\Omega| |T^c| < 1$ (with T^c being the complement of T) there exists a linear operator \mathbf{Q} and a constant p such that

$$\|f - \mathbf{Q}[r]\| \leq p \|n\|, \quad (12)$$

where $p \leq \left(1 - \sqrt{|T^c| |\Omega|}\right)^{-1}$. Specifically, the reconstruction operator \mathbf{Q} is given by

$$\mathbf{Q} = (\mathbf{I} - \mathbf{P}_T \mathbf{P}_\Omega)^{-1} = \sum_{k=0}^{\infty} (\mathbf{P}_T \mathbf{P}_\Omega)^k. \quad (13)$$

Moreover, the resulting solution is unique, and it can be approximated by truncating the Neumann series in (13) at some finite $k = N$.

An addition impetus to the theory of compressive sampling has been given in [7, 8, 9] via introducing the concept of two orthonormal bases Φ and Ψ of \mathbb{R}^n , which are used for sampling and signal representation, respectively. Moreover, central to the modern theory of compressive sampling are the notions of

- **Sparsity**, in the sense that signals can be represented by a relatively small number of non-zero coefficients in a properly chosen Ψ , and
- **Incoherence**, which represents the duality between the sampling Φ and representing Ψ domains, where the coherency

$$\mu(\Phi, \Psi) = \sqrt{n} \max |\Phi^T \Psi| \quad (14)$$

remains low. Note that here n is the number of vectors in the basis.

In its typical setting, compressive sampling refers to the case when an n -dimensional signals f has to be recovered from its m measurements

$$y_k = \langle f, \phi_k \rangle, \quad k \in M \subset \{1, \dots, n\} \quad (15)$$

where $m = \#M$ and $m < n$. Let Φ_M be the $n \times m$ matrix whose columns are formed by those ϕ_k for which $k \in M$. Then, assuming that the signal f can be represented as $f = \Psi c$, for some coefficient vector $c \in \mathbb{R}^n$, the reconstruction is carried out via solving

$$\min_c \|c\|_1 = \sum_{k=1}^n |c_k|, \quad s.t. \quad \Phi_M^T \Psi c = y, \quad (16)$$

where $y \in \mathbb{R}^m$ stands the vector of m measurements in (15).

The above problem can be solved by means of linear programming which, among all solutions obeying the measurement constraint $\Phi_M^T \Psi c = y$, picks the one that has the *sparsest* representation in the domain of Ψ as measured by the ℓ_1 -norm of c . Moreover, Theorem 1.2 derived in [10] defines a bound on the number of measurements m

$$m \geq C \mu^2(\Phi, \Psi) S \log n \quad (17)$$

for which *perfect* recovery is possible. Note that in (17), C is a constant, while S denotes the number of non-zero elements in $\Psi^T f$.

2.2 Reconstruction from Partial Derivatives

In the case when only partial derivatives of a signal of interest are available, the sampling operator of compressive sampling becomes the kernel of a derivative operator. In particular, in the 2-D case, we are given the measurements of $F_x = \partial F / \partial x$ and $F_y = \partial F / \partial y$. At this point, there are two possibilities to find F . The first would be to define Φ to be a discretized version of the 1st-order derivative operator. This choice, however, could result in relatively large values of the coherency $\mu(\Phi, \Psi)$ for the case when Ψ is a wavelet orthobasis (which is the choice in the present study). This would, in turn, increase the bound in (17), which could be unacceptable for practical considerations. On the other hand, one can define Φ to be the Dirac comb (i.e., $\Phi = \mathbf{I}$). In this case, the partial derivatives can be recovered first, followed by integrating the latter using (7).

To proceed with the second of the above-mentioned possibilities, we turn to a discrete setup in which F , F_x and F_y are considered to be $n \times n$ matrices. In this case, the maximal possible number of measurements is equal to $2n^2$, and hence $M \subset \{1, 2, \dots, 2n^2\}$. Specifically, we are interested in the case when $m = \#M < n^2$.

In 2-D, the partial derivatives F_x and F_y can be approximated according to

$$\begin{aligned} F_x &\simeq FD \\ F_y &\simeq D^T F, \end{aligned} \quad (18)$$

where D is 2-D difference matrix

$$D = \begin{bmatrix} 1 & 0 & \dots & 0 & 0 \\ -1 & 1 & \dots & 0 & 0 \\ 0 & -1 & \dots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 & 0 \\ 0 & 0 & \dots & -1 & 1 \end{bmatrix} \quad (19)$$

(Note that the last column of D defines the boundary condition.)

For the sake of notational simplicity, let $\Phi^\otimes = \Phi \otimes \Phi$ and $\Psi^\otimes = \Psi \otimes \Psi$, where \otimes stands for the Kronecker matrix product. Moreover, since the sampling sets for the x - and y -derivatives may be in general different, we denote the corresponding sampling matrices by Φ_x^\otimes and Φ_y^\otimes , respectively. Hence, assuming that there exist coefficients c_x and c_y such that $\text{vec}(F_x) = \Psi^\otimes \text{vec}(c_x)$ and $\text{vec}(F_y) = \Psi^\otimes \text{vec}(c_y)$ (with vec denoting the operation of matrix concatenation), the measurement constraints of the DCS problem are defined as

$$\begin{aligned} \Phi_x^\otimes \Psi^\otimes \text{vec}(c_x) &= G_x \\ \Phi_y^\otimes \Psi^\otimes \text{vec}(c_y) &= G_y, \end{aligned} \quad (20)$$

where G_x and G_y are the vectors of measured derivatives. In what follows, the constraints in (20) will be referred to as *primary*.

On the other hand, the cross-derivative (*secondary*) constraints in (8) can now be expressed as

$$\nabla_x \underbrace{\{\Psi c_y \Psi^T\}}_{F_y} = \nabla_y \underbrace{\{\Psi c_x \Psi^T\}}_{F_x}. \quad (21)$$

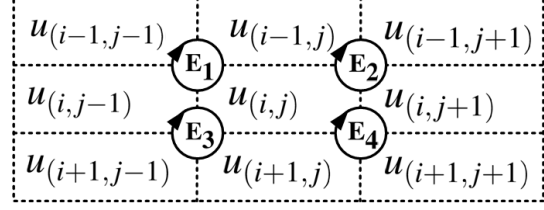


Figure 1: Circular integration paths involving the (i, j) pixel.

Using some standard rules of the matrix calculus [12], the above equation can be rewritten as

$$\left[\underbrace{(D^T \Psi) \otimes \Psi}_{\mathbf{B}_x} \quad - \underbrace{\Psi \otimes (D^T \Psi)}_{\mathbf{B}_y} \right] \underbrace{\begin{bmatrix} \text{vec}(c_x) \\ \text{vec}(c_y) \end{bmatrix}}_{\mathbf{c}} = 0 \quad (22)$$

The matrix $\mathbf{B} = [\mathbf{B}_x, -\mathbf{B}_y]$ is a full (row) rank matrix, whose condition number is approximately equal to $5n/4$. In the DCS formulation, this matrix of secondary (cross-derivative) constraints is combined with the primary constraints to result in the following optimization problem

$$\min_{\mathbf{c}} \|\mathbf{c}\|_1 = \sum_k |c_x(k)| + \sum_k |c_y(k)|, \quad (23)$$

subject to

$$\begin{bmatrix} \Phi_x^\otimes \Psi^\otimes & \mathbf{0} \\ \mathbf{0} & \Phi_y^\otimes \Psi^\otimes \\ \Phi_c^\otimes \mathbf{B}_x & -\Phi_c^\otimes \mathbf{B}_y \end{bmatrix} \mathbf{c} = \begin{bmatrix} G_x \\ G_y \end{bmatrix}$$

where Φ_c^\otimes denotes a sub-sampling operator which removes from the secondary constraints (21) those which are linearly dependent on the primary constraints (20) (see below).

Finally, having estimated the partial derivatives F_x and F_y as $\Psi^\otimes c_x$ and $\Psi^\otimes c_y$, respectively, we recover the function (phase) F as a solution to (7).

3. RECOVERING PRIMARY CONSTRAINTS FROM SECONDARY CONSTRAINTS

Since the gradient ∇F is a potential field, its integral over any closed path in \mathbb{R}^2 should be equal to zero, namely

$$\oint \nabla F(s) ds = 0. \quad (24)$$

Moreover, in the discrete case, the shortest of such paths connects each 4-pixels neighborhood, as shown in Fig. 1. The shown integration paths result in the following set of equations

$$\begin{aligned} \mathbf{q}_1 : F_y(i, j) &= F_y(i+1, j) + F_x(i, j) - F_x(i, j+1) \\ \mathbf{q}_2 : F_y(i, j) &= F_y(i-1, j) + F_x(i-1, j) - F_x(i-1, j) \\ \mathbf{q}_1 : F_x(i, j) &= F_x(i, j+1) + F_y(i, j) - F_y(i+1, j) \\ \mathbf{q}_3 : F_x(i, j) &= F_x(i, j-1) + F_y(i+1, j-1) - F_y(i, j-1). \end{aligned} \quad (25)$$

The constraint equations in (25) can be used to enlarge the set of primary constraints through recovering some of unknown values of F_x and F_y from those of their known neighbors. This procedure can be implemented using Algorithm 1 provided below.

Algorithm 1 Optimal recovery of primary constraint

```
while A pixel can be recovered do
  for  $i, j = 1$  to  $n$  do
    if all 3 elements in  $\mathbf{q}_1$  &  $\mathbf{q}_2$  are known then
       $F_y(i, j) = (\mathbf{q}_1 + \mathbf{q}_2)/2$ 
    else if all 3 elements in  $\mathbf{q}_1$  are known then
       $F_y(i, j) = \mathbf{q}_1$ 
    else if all 3 elements in  $\mathbf{q}_2$  are known then
       $F_y(i, j) = \mathbf{q}_2$ 
    end if
    if all 3 elements in  $\mathbf{q}_1$  &  $\mathbf{q}_3$  are known then
       $F_x(i, j) = (\mathbf{q}_1 + \mathbf{q}_3)/2$ 
    else if all 3 elements in  $\mathbf{q}_1$  are known then
       $F_x(i, j) = \mathbf{q}_1$ 
    else if all 3 elements in  $\mathbf{q}_3$  are known then
       $F_x(i, j) = \mathbf{q}_3$ 
    end if
  end for
end while
```

Algorithm 2 Elimination of dependent rows of \mathbf{B}

```
for  $i, j = 1$  to  $n$  do
  if  $F_y(i, j)$ : known AND  $F_x(i, j)$ : known then
    if  $F_y(i+1, j)$ : known OR  $F_x(i, j+1)$ : known then
       $\mathbf{B}(ij, :) = []$ ;
    end if
  else if  $F_y(i, j)$ : known OR  $F_x(i, j)$ : known then
    if  $F_y(i+1, j)$ : known AND  $F_x(i, j+1)$ : known then
       $\mathbf{B}(ij, :) = []$ ;
    end if
  else if  $F_y(i, j)$ : unknown AND  $F_x(i, j)$ : unknown then
    if  $F_y(i+1, j)$ : known AND  $F_x(i, j+1)$ : known then
       $\mathbf{B}(ij, :) = []$ ;
    end if
  end if
end for
```

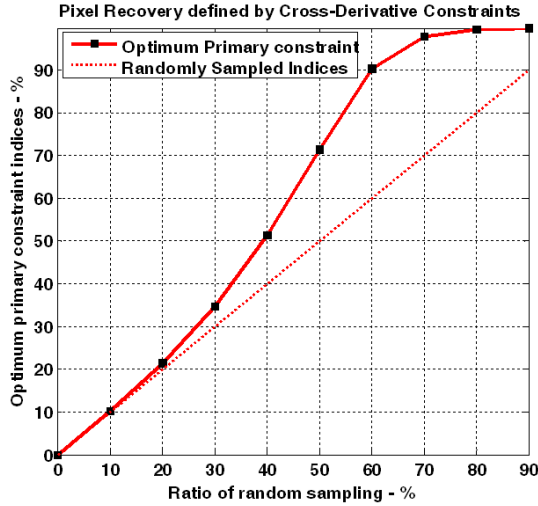


Figure 2: Recovering the indices of primary constraint for 256×256 test images.

Performing Algorithm 1 is a critical step as it maximizes the cardinality of the set of primary constraints, thereby improving the overall probability of recovering the true gradient. Typically, five iterations of the algorithm are sufficient to complete the task. Fig. 2 provides a quantitative characteristics of the algorithm which have been averaged over a number of 256×256 test images.

Finally, in order to exclude any linear dependency between the primary and secondary constraints, the matrix Φ_C^\otimes in (23) should be identified. However, in practice, instead of finding the matrix, we simply remove the rows of \mathbf{B} which are linearly dependent on the primary constraints. This can be done using Algorithm 2.

It should be pointed out that while Algorithm 1 maximizes the number of primary constraints, Algorithm 2 guarantees that the constraint matrix in (23) is of full row rank. After the execution of both algorithms, the problem of recovering the (sparse) representation coefficients \mathbf{c} can be solved by linear programming. However, in the 2-D case, such a solution could be rather impractical considering the size of

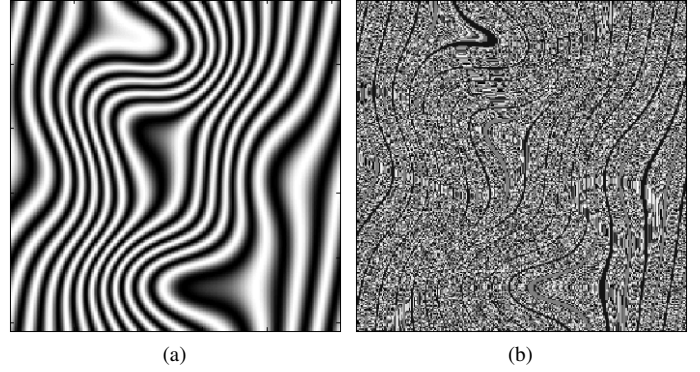


Figure 3: (a) Original phase; (b) Wrapped phase.

signals under consideration. To alleviate the computational burden, in the current work, the sparse solutions have been found using the algorithm detailed in [13].

4. RESULTS

We demonstrate the performance of our algorithm using a fringe pattern from the fringe phase data [14]. The original phase and its wrapped version are shown in Figure 3(a) and 3(b), respectively. Throughout the experimental study, Ψ was defined to be an orthogonal basis matrix corresponding to the nearly symmetric wavelet of I. Daubechies having six vanishing moments.

In our first numerical experiment, we compared the performance of the DCS algorithm with that of the standard CS method. Note that the latter can be obtained from the former by simply discarding the cross-derivative constraints (21). Through analyzing the field of residuals corresponding to the estimated phase gradient (see Fig. 4(b)), 23.76% of the total number of gradient samples were dismissed as unreliable. Fig. 4(a) shows the unwrapped phase estimated using the DCS algorithm. The mean-squared error (MSE) of the estimation was found to be 0.116%. As a comparison, the same solution was computed using the standard CS, whose MSE was found to be equal to 2.35%. Thus, one can see

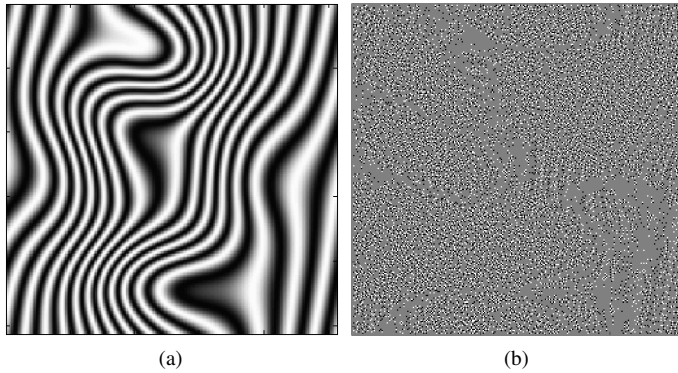


Figure 4: (a) Original phase estimated by the DCS method; (b) Residual points.

that augmenting the primary constraints of CS by the cross-derivative constraints results in substantial reduction in the level of MSE.

In our second experiment, we fixed the MSE of 2.35% as obtained by the standard CS above, and further reduced (via random exclusion) the number of available data points to find the percentage for which the DCS method would result in the same error rate. This percentage was found to be equal to 60%. As a comparison, for the same number of excluded data points, the standard CS resulted in MSE of 8.45%.

5. CONCLUSION

The main idea of derivative compressive sampling is to augment the constraints of the standard CS by some additional constraints related to the properties of the gradient as a potential field. In this case, the reconstruction algorithm resolves the ambiguity of “too few samples” not by only looking for a solution of maximal sparseness in the domain of Ψ , but also a solution that complies with the properties of a gradient field.

The proposed DCS algorithm is performed in two stages: first, the partial derivatives of a signal of interest are recovered from their sub-critically sampled measurements, followed by integrating the estimated derivatives via solving a Poisson equation. It should be noted, however, that it is possible to get rid of the second stage via defining the sampling system Φ to be a discretized version of the 1st-order derivative operator. In this case, the resulting CS problem should be capable of directly recovering the original signal. Unfortunately, this solution seems to be applicable only for recovering relatively smooth signals. This is because the derivative-based Φ can be incoherent only with bases of smooth (slow varying) functions, thereby ruling out the use of wavelets and the relatives thereof.

The current research results leave open a lot of exciting theoretical questions (e.g., as to what other constraints could be incorporated into the problem of CS). Moreover, it is still to be proved what bases could be considered to be optimal for representing the gradients of natural scenes [15]. Computation efficiency of DCS is another issue that should not be overseen when practical applications of DCS are of concern [16].

REFERENCES

- [1] J. D. Jackson, A. J. Yezzi, and S. Soatto, “Joint priors for variational shape and appearance modeling,” in *IEEE Conference on Computer Vision and Pattern Recognition, 2007. CVPR '07*, June 17-22. 2007, pp. 1–7.
- [2] D. C. Ghiglia, M. D. Pritt, *Two-dimensional phase unwrapping: Theory, algorithms, and software*. New York: Wiley-Interscience, 1998.
- [3] O. Michailovich and A. Tannenbaum, “A fast approximation of smooth functions from samples of partial derivatives with application to phase unwrapping,” *Signal Processing*, Vol. 88, pp. 358–374, 2008.
- [4] O. Michailovich and D. Adam, “Phase unwrapping for 2-D blind deconvolution of ultrasound images,” *IEEE Trans. Med. Imag.*, vol. 23, No. 1, pp. 7–25, Jan. 2004.
- [5] A. V. Oppenheim, R. W. Schaffer, *Discrete time signal processing*. London: Prentice Hall, 1989.
- [6] D. L. Donoho and P. B. Stark, “Uncertainty principles and signal recovery,” *SIAM J. Appl. Math.*, vol. 49, no. 3, pp. 906–931, June 1989.
- [7] E. Candes, J. Romberg and T. Tao, “Robust uncertainty principles: Exact signal reconstruction from highly incomplete frequency information,” *IEEE Trans. Inform. Theory*, vol. 52, no. 2, pp. 489–509, Feb. 2006.
- [8] E. Candes and T. Tao, “Near-optimal signal recovery from random projections: Universal encoding strategies?,” *IEEE Trans. Inform. Theory*, vol. 52, no. 12, pp. 4506–4525, Dec. 2006.
- [9] D. L. Donoho, “Compressed sensing,” *IEEE Trans. Inform. Theory*, vol. 52, no. 4, Apr. 2006.
- [10] E. Candes and J. Romberg, “Sparsity and incoherence in compressive sampling,” *Inverse Problems*, vol. 23, no. 3, pp. 969–985, 2007.
- [11] E. Candes and T. Tao, “Decoding by linear programming,” *IEEE Trans. Inform. Theory*, vol. 51, no. 12, Dec. 2005.
- [12] J. W. Brewer, “Kronecker products and matrix calculus in system theory,” *IEEE Trans. Circuits and Sys.*, vol. 25, no. 9, Sep. 1978.
- [13] E. V. D. Berg and M. P. Frielander, “Probing the pareto frontier for basis pursuit solution,” *Technical Report TR-2008-01, Department of CS, Univ. of British Columbia*, May. 2008.
- [14] J. A. Quiroga, D. Crespo and J. A. Gomez-Pedrero “XtremeFringe®: state-of-the-art software for automatic processing of fringe patterns,” *Optical Measurment Sys. for Industrial Inspection V, Proc. of SPIE*, vol. 6616 66163Y-1, 2007.
- [15] M. Aharon, M. Elad, and A.M. Bruckstein, “The K-SVD: An algorithm for designing of overcomplete dictionaries for sparse representation,” *IEEE Trans. On Signal Processing*, Vol. 54, no. 11, pp. 4311-4322, Nov. 2006.
- [16] D. L. Donoho, Y. Tsaig, I. Drori and J. I. Starck, “Sparse solution of underdetermined linear equations by stagewise orthogonal matching pursuit,” *Technical Report, Univ. of Stanford*, 2006.